

*It's a poor sort of memory that only works backward.*  
—Lewis Carroll (1832–1898)

All the Boolean and arithmetic chips that we built in chapters 1 and 2 were *combinational*. Combinational chips compute functions that depend solely on *combinations* of their input values. These relatively simple chips provide many important processing functions (like the ALU), but they cannot *maintain state*. Since computers must be able to not only compute values but also store and recall values, they must be equipped with memory elements that can preserve data over time. These memory elements are built from *sequential chips*.

The implementation of memory elements is an intricate art involving synchronization, clocking, and feedback loops. Conveniently, most of this complexity can be embedded in the operating logic of very low-level sequential gates called *flip-flops*. Using these flip-flops as elementary building blocks, we will specify and build all the memory devices employed by typical modern computers, from binary cells to registers to memory banks and counters. This effort will complete the construction of the chip set needed to build an entire computer—a challenge that we take up in the chapter 5.

Following a brief overview of clocks and flip-flops, the Background section introduces all the memory chips that we will build on top of them. The next two sections describe the chips Specification and Implementation, respectively. As usual, all the chips mentioned in the chapter can be built and tested using the hardware simulator supplied with the book, following the instructions given in the final Project section.

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### 3.1 Background

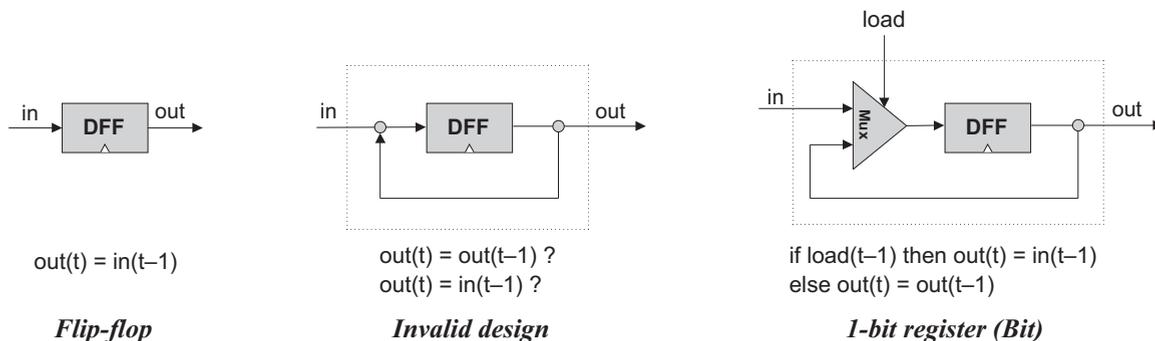
The act of “remembering something” is inherently time-dependent: You remember *now* what has been committed to memory *before*. Thus, in order to build chips that “remember” information, we must first develop some standard means for representing the progression of time.

**The Clock** In most computers, the passage of time is represented by a master clock that delivers a continuous train of alternating signals. The exact hardware implementation is usually based on an oscillator that alternates continuously between two phases labeled 0–1, *low-high*, *tick-tock*, etc. The elapsed time between the beginning of a “tick” and the end of the subsequent “tock” is called *cycle*, and each clock cycle is taken to model one discrete time unit. The current clock phase (*tick* or *tock*) is represented by a binary signal. Using the hardware’s circuitry, this signal is simultaneously broadcast to every sequential chip throughout the computer platform.

**Flip-Flops** The most elementary sequential element in the computer is a device called a *flip-flop*, of which there are several variants. In this book we use a variant called a *data flip-flop*, or DFF, whose interface consists of a single-bit data input and a single-bit data output. In addition, the DFF has a *clock* input that continuously changes according to the master clock’s signal. Taken together, the data and the clock inputs enable the DFF to implement the time-based behavior  $out(t) = in(t - 1)$ , where *in* and *out* are the gate’s input and output values and *t* is the current clock cycle. In other words, the DFF simply outputs the input value from the previous time unit.

As we now show, this elementary behavior can form the basis of all the hardware devices that computers use to *maintain state*, from binary cells to registers to arbitrarily large random access memory (RAM) units.

**Registers** A *register* is a storage device that can “store,” or “remember,” a value over time, implementing the classical storage behavior  $out(t) = out(t - 1)$ . A DFF, on the other hand, can only output its previous input, namely,  $out(t) = in(t - 1)$ . This suggests that a register can be implemented from a DFF by simply feeding the output of the latter back into its input, creating the device shown in the middle of figure 3.1. Presumably, the output of this device at any time *t* will echo its output at time *t* – 1, yielding the classical function expected from a storage unit.



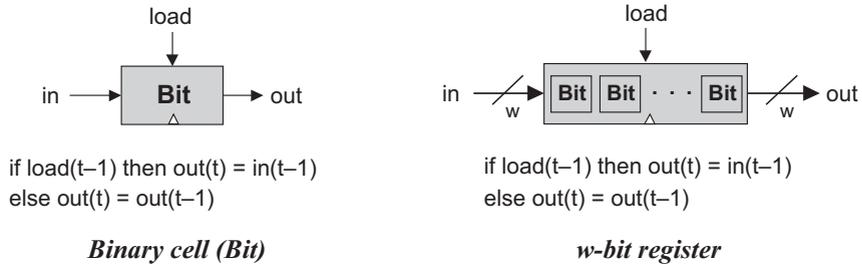
**Figure 3.1** From a DFF to a single-bit register. The small triangle represents the clock input. This icon is used to state that the marked chip, as well as the overall chip that encapsulates it, is time-dependent.

Well, not so. The device shown in the middle of figure 3.1 is invalid. First, it is not clear how we’ll ever be able to load this device with a new data value, since there are no means to tell the DFF when to draw its input from the *in* wire and when from the *out* wire. More generally, the rules of chip design dictate that internal pins must have a fan-in of 1, meaning that they can be fed from a single source only.

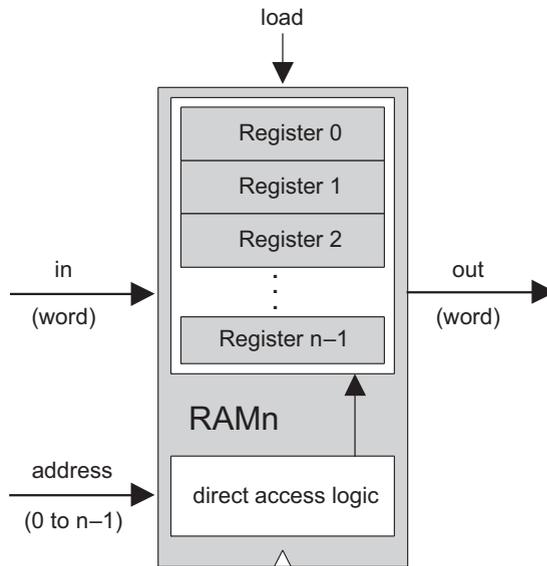
The good thing about this thought experiment is that it leads us to the correct and elegant solution shown in the right side of figure 3.1. In particular, a natural way to resolve our input ambiguity is to introduce a multiplexor into the design. Further, the “select bit” of this multiplexor can become the “load bit” of the overall register chip: If we want the register to start storing a new value, we can put this value in the *in* input and set the *load* bit to 1; if we want the register to keep storing its internal value until further notice, we can set the *load* bit to 0.

Once we have developed the basic mechanism for remembering a single bit over time, we can easily construct arbitrarily wide registers. This can be achieved by forming an array of as many single-bit registers as needed, creating a register that holds multi-bit values (figure 3.2). The basic design parameter of such a register is its *width*—the number of bits that it holds—e.g., 16, 32, or 64. The multi-bit contents of such registers are typically referred to as *words*.

**Memories** Once we have the basic ability to represent words, we can proceed to build memory banks of arbitrary length. As figure 3.3 shows, this can be done by stacking together many registers to form a *Random Access Memory* RAM unit. The term *random access memory* derives from the requirement that read/write operations



**Figure 3.2** From single-bit to multi-bit registers. A multi-bit register of width  $w$  can be constructed from an array of  $w$  1-bit chips. The operating functions of both chips is exactly the same, except that the “=” assignments are single-bit and multi-bit, respectively.



**Figure 3.3** RAM chip (conceptual). The width and length of the RAM can vary.

on a RAM should be able to access randomly chosen words, with no restrictions on the order in which they are accessed. That is to say, we require that any word in the memory—irrespective of its physical location—be accessed directly, in equal speed.

This requirement can be satisfied as follows. First, we assign each word in the  $n$ -register RAM a unique *address* (an integer between 0 to  $n - 1$ ), according to which it will be accessed. Second, in addition to building an array of  $n$  registers, we build a gate logic design that, given an address  $j$ , is capable of selecting the individual register whose address is  $j$ . Note however that the notion of an “address” is not an explicit part of the RAM design, since the registers are not “marked” with addresses in any physical sense. Rather, as we will see later, the chip is equipped with direct access logic that implements the notion of addressing using logical means.

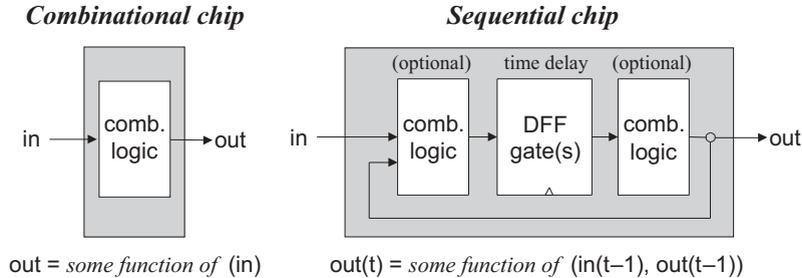
In sum, a classical RAM device accepts three inputs: a data input, an address input, and a load bit. The *address* specifies which RAM register should be accessed in the current time unit. In the case of a read operation ( $\text{load}=0$ ), the RAM’s output immediately emits the value of the selected register. In the case of a write operation ( $\text{load}=1$ ), the selected memory register commits to the input value in the next time unit, at which point the RAM’s output will start emitting it.

The basic design parameters of a RAM device are its data *width*—the width of each one of its words, and its *size*—the number of words in the RAM. Modern computers typically employ 32- or 64-bit-wide RAMs whose sizes are up to hundreds of millions.

**Counters** A counter is a sequential chip whose state is an integer number that increments every time unit, effecting the function  $out(t) = out(t - 1) + c$ , where  $c$  is typically 1. Counters play an important role in digital architectures. For example, a typical CPU includes a *program counter* whose output is interpreted as the address of the instruction that should be executed next in the current program.

A counter chip can be implemented by combining the input/output logic of a standard register with the combinatorial logic for adding a constant. Typically, the counter will have to be equipped with some additional functionality, such as possibilities for resetting the count to zero, loading a new counting base, or decrementing instead of incrementing.

**Time Matters** All the chips described so far in this chapter are *sequential*. Simply stated, a sequential chip is a chip that embeds one or more DFF gates, either directly or indirectly. Functionally speaking, the DFF gates endow sequential chips with the ability to either maintain state (as in memory units) or operate on state (as in



**Figure 3.4** Combinational versus sequential logic (in and out stand for one or more input and output variables). Sequential chips always consist of a layer of DFFs sandwiched between optional combinational logic layers.

counters). Technically speaking, this is done by forming feedback loops inside the sequential chip (see figure 3.4). In combinational chips, where time is neither modeled nor recognized, the introduction of feedback loops is problematic: The output would depend on the input, which itself would depend on the output, and thus the output would depend on itself. On the other hand, there is no difficulty in feeding the output of a sequential chip back into itself, since the DFFs introduce an inherent time delay: The output at time  $t$  does not depend on itself, but rather on the output at time  $t - 1$ . This property guards against the uncontrolled “data races” that would occur in combinational chips with feedback loops.

Recall that the outputs of combinational chips change when their inputs change, irrespective of time. In contrast, the inclusion of the DFFs in the sequential architecture ensures that their outputs change only at the point of transition from one clock cycle to the next, and not within the cycle itself. In fact, we allow sequential chips to be in unstable states *during* clock cycles, requiring only that at the beginning of the next cycle they output correct values.

This “discretization” of the sequential chips’ outputs has an important side effect: It can be used to synchronize the overall computer architecture. To illustrate, suppose we instruct the arithmetic logic unit (ALU) to compute  $x + y$  where  $x$  is the value of a nearby register and  $y$  is the value of a remote RAM register. Because of various physical constraints (distance, resistance, interference, random noise, etc.) the electric signals representing  $x$  and  $y$  will likely arrive at the ALU at different times. However, being a *combinational chip*, the ALU is insensitive to the concept of time—it continuously adds up whichever data values happen to lodge in its inputs. Thus, it will take some time before the ALU’s output stabilizes to the correct  $x + y$  result. Until then, the ALU will generate garbage.

How can we overcome this difficulty? Well, since the output of the ALU is always routed to some sort of a sequential chip (a register, a RAM location, etc.), *we don't really care*. All we have to do is ensure, when we build the computer's clock, that the length of the clock cycle will be slightly longer than the time it takes a bit to travel the longest distance from one chip in the architecture to another. This way, we are guaranteed that by the time the sequential chip updates its state (at the beginning of the next clock cycle), the inputs that it receives from the ALU will be valid. This, in a nutshell, is the trick that synchronizes a set of stand-alone hardware components into a well-coordinated system, as we shall see in chapter 5.

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## 3.2 Specification

This section specifies a hierarchy of sequential chips:

- Data-flip-flops (DFFs)
- Registers (based on DFFs)
- Memory banks (based on registers)
- Counter chips (also based on registers)

### 3.2.1 Data-Flip-Flop

The most elementary sequential device that we present—the basic component from which all memory elements will be designed—is the *data flip-flop* gate. A DFF gate has a single-bit input and a single-bit output, as follows:



```

Chip name: DFF
Inputs:    in
Outputs:  out
Function: out(t)=in(t-1)
Comment:  This clocked gate has a built-in
              implementation and thus there is
              no need to implement it.
  
```

Like Nand gates, DFF gates enter our computer architecture at a very low level. Specifically, all the sequential chips in the computer (registers, memory, and

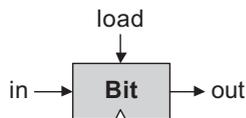
counters) are based on numerous DFF gates. All these DFFs are connected to the same master clock, forming a huge distributed “chorus line.” At the beginning of each clock cycle, the outputs of *all* the DFFs in the computer commit to their inputs during the previous time unit. At all other times, the DFFs are “latched,” meaning that changes in their inputs have no immediate effect on their outputs. This conduction operation effects any one of the millions of DFF gates that make up the system, about a billion times per second (depending on the computer’s clock frequency).

Hardware implementations achieve this time dependency by simultaneously feeding the master clock signal to all the DFF gates in the platform. Hardware simulators emulate the same effect in software. As far as the computer architect is concerned, the end result is the same: The inclusion of a DFF gate in the design of any chip ensures that the overall chip, as well as all the chips up the hardware hierarchy that depend on it, will be inherently time-dependent. These chips are called *sequential*, by definition.

The physical implementation of a DFF is an intricate task, and is based on connecting several elementary logic gates using feedback loops (one classic design is based on Nand gates alone). In this book we have chosen to abstract away this complexity, treating DFFs as primitive building blocks. Thus, our hardware simulator provides a built-in DFF implementation that can be readily used by other chips.

### 3.2.2 Registers

A single-bit register, which we call *Bit*, or *binary cell*, is designed to store a single bit of information (0 or 1). The chip interface consists of an input pin that carries a data bit, a *load* pin that enables the cell for writes, and an output pin that emits the current state of the cell. The interface diagram and API of a binary cell are as follows:

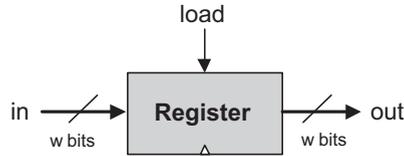


```

Chip name: Bit
Inputs:    in, load
Outputs:  out
Function:  If load(t-1) then out(t)=in(t-1)
               else out(t)=out(t-1)

```

The API of the Register chip is essentially the same, except that the input and output pins are designed to handle multi-bit values:



**Chip name:** Register  
**Inputs:** in[16], load  
**Outputs:** out[16]  
**Function:** If load( $t-1$ ) then out( $t$ )=in( $t-1$ )  
 else out( $t$ )=out( $t-1$ )  
**Comment:** "=" is a 16-bit operation.

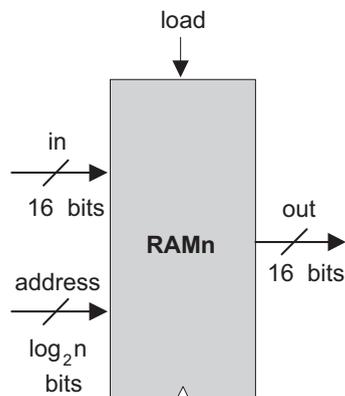
The Bit and Register chips have exactly the same read/write behavior:

*Read:* To read the contents of a register, we simply probe its output.

*Write:* To write a new data value  $d$  into a register, we put  $d$  in the in input and assert (set to 1) the load input. In the next clock cycle, the register commits to the new data value, and its output starts emitting  $d$ .

### 3.2.3 Memory

A direct-access memory unit, also called RAM, is an array of  $n$   $w$ -bit registers, equipped with direct access circuitry. The number of registers ( $n$ ) and the width of each register ( $w$ ) are called the memory's *size* and *width*, respectively. We will now set out to build a hierarchy of such memory devices, all 16 bits wide, but with varying sizes: RAM8, RAM64, RAM512, RAM4K, and RAM16K units. All these memory chips have precisely the same API, and thus we describe them in one parametric diagram:



**Chip name:** RAMn //  $n$  and  $k$  are listed below  
**Inputs:** in[16], address[ $k$ ], load  
**Outputs:** out[16]  
**Function:** out( $t$ )=RAM[address( $t$ )]( $t$ )  
 If load( $t-1$ ) then  
     RAM[address( $t-1$ )]( $t$ )=in( $t-1$ )  
**Comment:** "=" is a 16-bit operation.

The specific RAM chips needed for the Hack platform are:

Chip name	$n$	$k$
RAM8	8	3
RAM64	64	6
RAM512	512	9
RAM4K	4096	12
RAM16K	16384	14

*Read:* To read the contents of register number  $m$ , we put  $m$  in the `address` input. The RAM's direct-access logic will select register number  $m$ , which will then emit its output value to the RAM's output pin. This is a combinational operation, independent of the clock.

*Write:* To write a new data value  $d$  into register number  $m$ , we put  $m$  in the `address` input,  $d$  in the `in` input, and assert the `load` input bit. This causes the RAM's direct-access logic to select register number  $m$ , and the `load` bit to enable it. In the next clock cycle, the selected register will commit to the new value ( $d$ ), and the RAM's output will start emitting it.

### 3.2.4 Counter

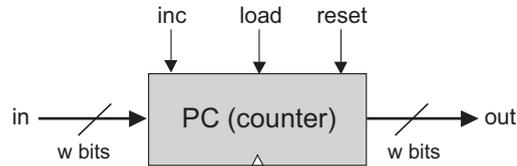
Although a *counter* is a stand-alone abstraction in its own right, it is convenient to motivate its specification by saying a few words about the context in which it is normally used. For example, consider a counter chip designed to contain the address of the instruction that the computer should fetch and execute next. In most cases, the counter has to simply increment itself by 1 in each clock cycle, thus causing the computer to fetch the next instruction in the program. In other cases, for example, in “jump to execute instruction number  $n$ ,” we want to be able to set the counter to  $n$ , then have it continue its default counting behavior with  $n + 1$ ,  $n + 2$ , and so forth. Finally, the program's execution can be restarted anytime by resetting the counter to 0, assuming that that's the address of the program's first instruction. In short, we need a loadable and resettable counter.

With that in mind, the interface of our Counter chip is similar to that of a register, except that it has two additional control bits labeled `reset` and `inc`. When `inc=1`, the counter increments its state in every clock cycle, emitting the value  $\text{out}(\tau) = \text{out}(\tau-1) + 1$ . If we want to reset the counter to 0, we assert the `reset` bit; if we want to initialize it to some other counting base  $d$ , we put  $d$  in the `in` input and assert the `load` bit. The details are given in the counter API, and an example of its operation is depicted in figure 3.5.

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## 3.3 Implementation

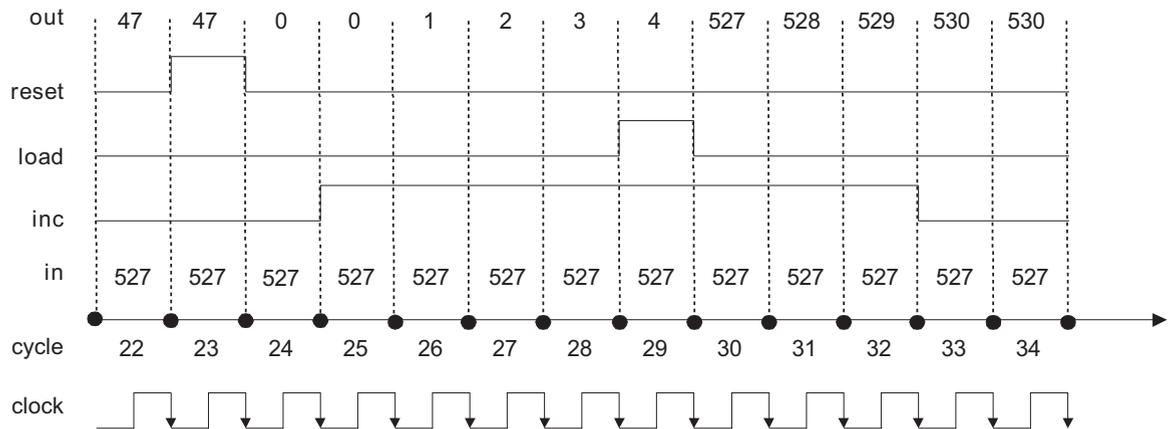
**Flip-Flop** DFF gates can be implemented from lower-level logic gates like those built in chapter 1. However, in this book we treat DFFs as primitive gates, and thus they can be used in hardware construction projects without worrying about their internal implementation.



```

Chip name: PC // 16-bit counter
Inputs:   in[16], inc, load, reset
Outputs: out[16]
Function: If reset(t-1) then out(t)=0
              else if load(t-1) then out(t)=in(t-1)
              else if inc(t-1) then out(t)=out(t-1)+1
              else out(t)=out(t-1)
Comment:  "=" is 16-bit assignment.
              "+" is 16-bit arithmetic addition.

```



We assume that we start tracking the counter in time unit 22, when its input and output happen to be 527 and 47, respectively. We also assume that the counter's control bits (reset, load, inc) start at 0—all arbitrary assumptions.

**Figure 3.5** Counter simulation. At time 23 a *reset* signal is issued, causing the counter to emit 0 in the following time unit. The 0 persists until an *inc* signal is issued at time 25, causing the counter to start incrementing, one time unit later. The counting continues until, at time 29, the load bit is asserted. Since the counter's input holds the number 527, the counter is reset to that value in the next time unit. Since inc is still asserted, the counter continues incrementing until time 33, when inc is de-asserted.

**1-Bit Register (Bit)** The implementation of this chip was given in figure 3.1.

**Register** The construction of a  $w$ -bit Register chip from 1-bit registers is straightforward. All we have to do is construct an array of  $w$  Bit gates and feed the register's load input to every one of them.

**8-Register Memory (RAM8)** An inspection of figure 3.3 may be useful here. To implement a RAM8 chip, we line up an array of eight registers. Next, we have to build combinational logic that, given a certain `address` value, takes the RAM8's `in` input and loads it into the selected register. In a similar fashion, we have to build combinational logic that, given a certain `address` value, selects the right register and pipes its out value to the RAM8's `out` output. Tip: This combinational logic was already implemented in chapter 1.

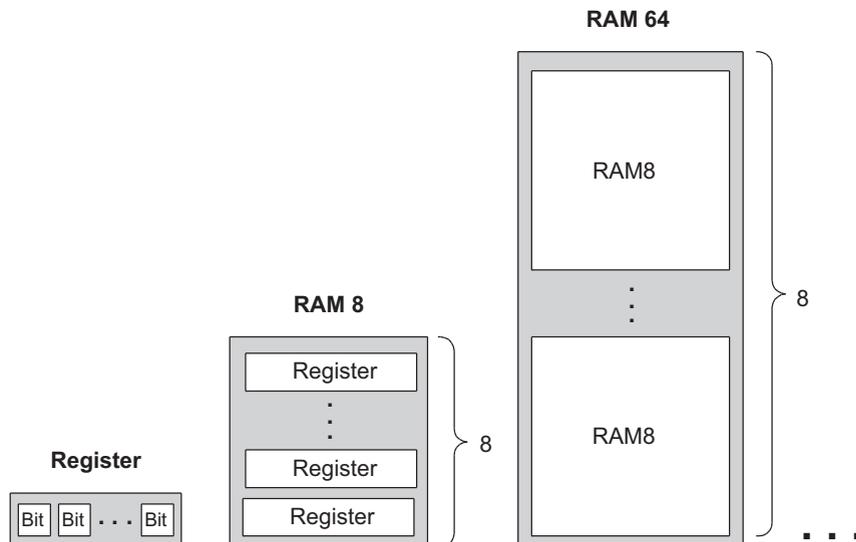
**$n$ -Register Memory** A memory bank of arbitrary length (a power of 2) can be built recursively from smaller memory units, all the way down to the single register level. This view is depicted in figure 3.6. Focusing on the right-hand side of the figure, we note that a 64-register RAM can be built from an array of eight 8-register RAM chips. To select a particular register from the RAM64 memory, we use a 6-bit address, say `xxxyyy`. The MSB `xxx` bits select one of the RAM8 chips, and the LSB `yyy` bits select one of the registers within the selected RAM8. The RAM64 chip should be equipped with logic circuits that effect this hierarchical addressing scheme.

**Counter** A  $w$ -bit counter consists of two main elements: a regular  $w$ -bit register, and combinational logic. The combinational logic is designed to (a) compute the counting function, and (b) put the counter in the right operating mode, as mandated by the values of its three control bits. Tip: Most of this logic was already built in chapter 2.

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### 3.4 Perspective

The cornerstone of all the memory systems described in this chapter is the flip-flop—a gate that we treated here as an atomic, or primitive, building block. The usual approach in hardware textbooks is to construct flip-flops from elementary combinatorial gates (e.g., Nand gates) using appropriate feedback loops. The standard con-



**Figure 3.6** Gradual construction of memory banks by recursive ascent. A  $w$ -bit register is an array of  $w$  binary cells, an 8-register RAM is an array of eight  $w$ -bit registers, a 64-register RAM is an array of eight RAM8 chips, and so on. Only three more similar construction steps are necessary to build a 16K RAM chip.

struction begins by building a simple (non-clocked) flip-flop that is bi-stable, namely, that can be set to be in one of two states. Then a clocked flip-flop is obtained by cascading two such simple flip-flops, the first being set when the clock *ticks* and the second when the clock *tocks*. This “master-slave” design endows the overall flip-flop with the desired clocked synchronization functionality.

These constructions are rather elaborate, requiring an understating of delicate issues like the effect of feedback loops on combinatorial circuits, as well as the implementation of clock cycles using a two-phase binary clock signal. In this book we have chosen to abstract away these low-level considerations by treating the flip-flop as an atomic gate. Readers who wish to explore the internal structure of flip-flop gates can find detailed descriptions in most logic design and computer architecture textbooks.

In closing, we should mention that memory devices of modern computers are not always constructed from standard flip-flops. Instead, modern memory chips are usually very carefully optimized, exploiting the unique physical properties of the underlying storage technology. Many such alternative technologies are available

today to computer designers; as usual, which technology to use is a cost-performance issue.

Aside from these low-level considerations, all the other chip constructions in this chapter—the registers and memory chips that were built on top of the flip-flop gates—were standard.

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### 3.5 Project

**Objective** Build all the chips described in the chapter. The only building blocks that you can use are primitive DFF gates, chips that you will build on top of them, and chips described in previous chapters.

**Resources** The only tool that you need for this project is the hardware simulator supplied with the book. All the chips should be implemented in the HDL language specified in appendix A. As usual, for each chip we supply a skeletal `.hdl` program with a missing implementation part, a `.tst` script file that tells the hardware simulator how to test it, and a `.cmp` compare file. Your job is to complete the missing implementation parts of the supplied `.hdl` programs.

**Contract** When loaded into the hardware simulator, your chip design (modified `.hdl` program), tested on the supplied `.tst` file, should produce the outputs listed in the supplied `.cmp` file. If that is not the case, the simulator will let you know.

**Tip** The Data Flip-Flop (DFF) gate is considered primitive and thus there is no need to build it: When the simulator encounters a DFF gate in an HDL program, it automatically invokes the built-in `tools/builtIn/DFF.hdl` implementation.

**The Directory Structure of This Project** When constructing RAM chips from smaller ones, we recommend using built-in versions of the latter. Otherwise, the simulator may run very slowly or even out of (real) memory space, since large RAM chips contain tens of thousands of lower-level chips, and all these chips are kept in memory (as software objects) by the simulator. For this reason, we have placed the `RAM512.hdl`, `RAM4K.hdl`, and `RAM16K.hdl` programs in a separate directory. This way, the recursive descent construction of the RAM4K and RAM16K chips stops with the RAM512 chip, whereas the lower-level chips from which the latter chip

is made are bound to be built-in (since the simulator does not find them in this directory).

**Steps** We recommend proceeding in the following order:

0. The hardware simulator needed for this project is available in the `tools` directory of the book's software suite.
1. Read appendix A, focusing on sections A.6 and A.7.
2. Go through the *hardware simulator tutorial*, focusing on parts IV and V.
3. Build and simulate all the chips specified in the `projects/03` directory.

